

Estimation of High Dimensional Covariance Matrices

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Interest for the estimation of covariance matrices

- Many applied estimation methods rely on the second-order statistics of the observed random vector process:

$$\mathbf{y}_i = \mathbf{\Sigma}^{\frac{1}{2}} \mathbf{x}_i, i = 1, \dots, n \text{ and } \mathbf{y}_i \in \mathbb{C}^N$$

$$\mathbf{Y} = [\mathbf{y}_1, \dots, \mathbf{y}_n], \mathbf{X} = [\mathbf{x}_1, \dots, \mathbf{x}_n] \Rightarrow \mathbf{Y} = \mathbf{\Sigma}^{\frac{1}{2}} \mathbf{X}$$

Examples:

- ▶ Array processing: Estimation of direction of arrivals
- ▶ Inference of performance metrics (Capacity or SINR)
- Possible asymptotic regimes:
 - ▶ Number of observations is large as compared to the dimension of the received vector, $n \gg N$
 - ▶ Number of observations is of the same order of magnitude as the dimension of the received vector $n \propto N$. In the sequel, This regime will be referred to as :
The Regime of Large Random Matrices

Shortcomings of the conventional estimation techniques

- Consider n observations y_1, \dots, y_n independent and identically distributed where

$$\mathbf{y}_i = \mathbf{\Sigma}^{\frac{1}{2}} \mathbf{x}_i, i = 1, \dots, n \text{ and } \mathbf{y}_i \in \mathbb{C}^N$$

with the entries of \mathbf{x}_i are zero mean independent random variables such that $\mathbb{E}[\mathbf{x}_i \mathbf{x}_i^H] = \mathbf{I}_N$. The sample covariance estimate $\hat{\mathbf{\Sigma}}$ of $\mathbf{\Sigma}$ is given by:

$$\hat{\mathbf{\Sigma}} = \sum_{i=1}^n \mathbf{y}_i \mathbf{y}_i^H = \frac{\mathbf{Y} \mathbf{Y}^H}{n}$$

- If $n \gg N$, we know that $\|\hat{\mathbf{\Sigma}} - \mathbf{\Sigma}\| \xrightarrow[n \rightarrow \infty]{a.s.} 0..$
 - If $n \propto N$ ($\frac{N}{n} \rightarrow c$), $\|\hat{\mathbf{\Sigma}} - \mathbf{\Sigma}\|$ **does not converge** to zero as N and n grow at infinity.
- \Rightarrow Need to develop new techniques adapted to this regime

Theory of Large Random Matrices

- The theory of large random matrices offers powerful tools to deal with the regime N and n growing to infinity at the same pace.
- Brief History of the use of this theory:
 - ▶ First results of random matrix theory were motivated by nuclear physics (Wigner 1950),
 - ▶ This theory has been firstly applied to the field of wireless communication in the mid-90 with the pioneering works of Telatar and Foschini.
- ⇒ New approaches to analyze the performance of wireless communication systems.
- ▶ In 2007, X. Mestre uses this theory to propose new consistent estimation methods,
- ⇒ Practical interest of this theory to propose new efficient algorithms,

Results from the theory of large random matrices

- First results concern Wigner matrices: symmetric random matrices whose diagonal entries are 0 and whose upper-triangle entries are i.i.d taking the values ± 1 .
- Let $\mathbf{W}_N \in \mathbb{C}^{N \times N}$ be a Wigner matrix.
Then, as $N \rightarrow +\infty$, the distribution of the eigenvalues of $\frac{1}{\sqrt{N}}\mathbf{W}_N$ converges to the semi-circle law whose density is:

$$w(x) = \begin{cases} \frac{1}{2\pi} \sqrt{4 - x^2} & \text{if } |x| < 2. \\ 0 & \text{if } |x| > 2. \end{cases}$$

- This result was proven using the Moments method.

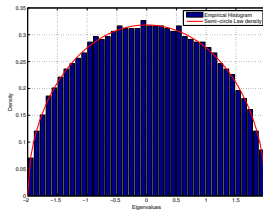


Figure: Histogram and Limit law ($N = 2000$)

Results from the theory of large random matrices

Generalization to singular values of rectangular matrices

- Limit Law of the eigenvalue distribution of

$$\mathbf{Y} = \mathbf{Y}_0 + \mathbf{X}\mathbf{T}\mathbf{X}^H$$

where \mathbf{T} is a real diagonal matrix independent of \mathbf{X} and \mathbf{Y}_0 is a deterministic hermitian matrix.

- If $\mathbf{Y}_0 \neq 0$ or $\mathbf{T} \neq \mathbf{I}$, no closed form expression exists for the limiting spectrum.
- Instead, it can be characterized using the Stieltjes transform method.

Stieltjes Transform of a random variable

- Stieltjes transform: Let X be a real-valued random variable with distribution $F_X(\cdot)$ (with support \mathcal{S}). The stieltjes transform is the Cauchy transform of F_X defined for complex arguments outside the support \mathcal{S} as:

$$m_X(z) = \mathbb{E} \left[\frac{1}{X - z} \right] = \int_{-\infty}^{+\infty} \frac{1}{\lambda - z} dF_X(\lambda)$$

- The Stieltjes transform is to the cauchy transform as the characteristic function is to the Fourier transform.
- Given $m_X(z)$, the probability density function of X can be retrieved as :

$$f_X(\lambda) = \lim_{\omega \rightarrow 0^+} \frac{1}{\pi} \Im [m_X(\lambda + j\omega)]$$

Stieltjes Transform of the covariance matrix

Stieltjes transform of a matrix is the Stieltjes transform of the empirical measure of its eigenvalues:

- Consider the model:

$$\mathbf{Y} = \mathbf{\Sigma}^{\frac{1}{2}} \mathbf{X}$$

- Let $\lambda_1, \dots, \lambda_N$ be the eigenvalues of $\frac{1}{N} \mathbf{Y} \mathbf{Y}^H$. Denote by $dF^{\frac{1}{N} \mathbf{Y} \mathbf{Y}^H}$ the empirical measure for $\frac{1}{N} \mathbf{Y} \mathbf{Y}^H$:

$$dF^{\frac{1}{N} \mathbf{Y} \mathbf{Y}^H} = \frac{1}{N} \sum_{k=1}^N \delta_{\lambda_k}$$

- The Stieltjes transform for $dF^{\frac{1}{N} \mathbf{Y} \mathbf{Y}^H}$ is therefore given by:

$$\hat{m}_{\frac{1}{N} \mathbf{Y} \mathbf{Y}^H}(z) = \frac{1}{N} \sum_{k=1}^N \frac{1}{\lambda_k - z} = \frac{1}{N} \text{tr} \left(\frac{1}{N} \mathbf{Y} \mathbf{Y}^H - z \mathbf{I}_N \right)^{-1}$$

⇒ The study of the Stieltjes transform of the empirical measure amounts to studying the behaviour of the trace of the resolvent matrix:

$$\mathbf{Q}(z) \triangleq \left(\frac{1}{N} \mathbf{Y} \mathbf{Y}^H - z \mathbf{I}_N \right)^{-1}$$

Stieltjes transform of covariance matrices

- Let $\alpha_1, \dots, \alpha_N$ be the eigenvalues of Σ .

In the regime of large random matrices when $N, n \rightarrow +\infty$ with $\frac{N}{n} \rightarrow c$, the stieltjes transform of the empirical measure converges almost surely to $\bar{m}(z)$ where $\bar{m}(z)$ is the unique solution to:

$$\bar{m} = \frac{1}{N} \sum_{i=1}^N \frac{1}{\alpha_i(1 - c - cz\bar{m}) - z}.$$

- Let $\theta = \frac{1}{N} \sum_{k=1}^N f(\alpha_k)$, where f is a certain function. The conventional empirical estimate of θ is

$$\hat{\theta}_N = \frac{1}{N} \sum_{i=1}^N f(\lambda_i)$$

With the above result on hand, we can study the behaviour of $\hat{\theta}_N$ since,

$$\frac{1}{N} \sum_{k=1}^N f(\lambda_i) = \frac{1}{2\pi j} \oint_{\mathcal{C}} f(z) \hat{m}_{\frac{1}{N}} \mathbf{Y} \mathbf{Y}^H(z) dz \rightarrow \frac{1}{2\pi j} \oint_{\mathcal{C}} f(z) \bar{m}(z) dz \neq \theta$$

Moment method for the estimation of eigenvalues

Work published in [IEEE-SP 2012]

- System model:

$$\mathbf{Y} = \mathbf{\Sigma}^{\frac{1}{2}} \mathbf{X}$$

where $\mathbf{\Sigma}$ has a finite number K of unknown eigenvalues β_1, \dots, β_K with unknown multiplicities N_1, \dots, N_K .

- We consider the problem of estimating the eigenvalues along with their multiplicities in the regime

$$N, n \rightarrow +\infty \text{ with } \frac{N}{n} \rightarrow c.$$

Moment-method for the estimation of eigenvalues

- We assume that Σ has K eigenvalues β_1, \dots, β_K with unknown multiplicities N_1, \dots, N_K ,
- The moment method is based on a consistent estimation of the moments

$$\theta_\ell \triangleq \sum_{k=1}^K \frac{N_k}{N} \beta_k^\ell$$

- Write θ_ℓ as function of the non-observable Stieltjes transform:

$$\theta_\ell = \sum_{i=1}^K \frac{N_i}{N} \beta_i^\ell = \frac{1}{2\pi j} \oint_C \omega^\ell m_\Sigma(\omega) d\omega$$

Moment-method for the estimation of eigenvalues

- The limit $\underline{m}_N(z)$ of the observable Stieltjes transform of matrix $\frac{1}{N}\mathbf{Y}^H\mathbf{Y}$ satisfies:

$$\underline{m}_N = -zm_\Sigma \left(-\frac{1}{\underline{m}} \right) \left(\frac{n}{N} m_\Sigma \left(-\frac{1}{\underline{m}_N(z)} \right) - \frac{N-n}{Nz} \right)$$

- Using the changing variable $\omega = -\frac{1}{\underline{m}_N(z)}$, we obtain,

$$\theta_k = \frac{(-1)^k}{2j\pi} \oint_{\mathcal{C}} \frac{zm'_N(z)}{\underline{m}_N^{k+1}(z)} \left(c\underline{m}_N(z) + \frac{1-c}{cz} \right).$$

Moment method for the estimation of the eigenvalues

- Recall that :

$$\theta_k = \frac{(-1)^k}{2j\pi} \oint_C \frac{z \underline{m}'_N(z)}{\underline{m}_N^{k+1}(z)} \left(c \underline{m}_N(z) + \frac{1-c}{cz} \right).$$

\Rightarrow Compute consistent estimates $\hat{\theta}_k$ of θ_k by substituting $\underline{m}_N(z)$ by the stieltjes transform corresponding to matrix $\frac{1}{N} \mathbf{Y}^H \mathbf{Y}$.

- Consider the following system of equations

$$\begin{cases} \sum_{i=1}^K x_i & = 1 \\ \sum_{i=1}^K x_i y_i^k & = \hat{\theta}_k \end{cases}$$

admits a unique solution $(\hat{c}_1, \dots, \hat{c}_K, \hat{\beta}_1, \dots, \hat{\beta}_K)$ which is a consistent estimator of $(\frac{N_1}{N}, \dots, \frac{N_K}{N})$ and $(\beta_1, \dots, \beta_K)$, i.e, for each $1 \leq \ell \leq K$,

$$\hat{c}_\ell - \frac{N_\ell}{N} \xrightarrow[n, N \rightarrow +\infty]{a.s.} 0 \text{ and } \hat{\beta}_\ell - \beta_\ell \xrightarrow[n, N \rightarrow +\infty]{a.s.} 0$$

Moment method for the estimation of the eigenvalues

Algorithm for solving to the obtained system of equations:

- Let $\mathbf{\Gamma} = \begin{bmatrix} \hat{\theta}_0 & \hat{\theta}_1 & \cdots & \hat{\theta}_{K-1} \\ \hat{\theta}_1 & \hat{\theta}_2 & \cdots & \hat{\theta}_K \\ \vdots & \ddots & \ddots & \vdots \\ \hat{\theta}_{K-1} & \hat{\theta}_K & \cdots & \hat{\theta}_{2K-2} \end{bmatrix}$ and $\mathbf{b} = [\hat{\theta}_L, \dots, \hat{\theta}_{2K-1}]^T$,

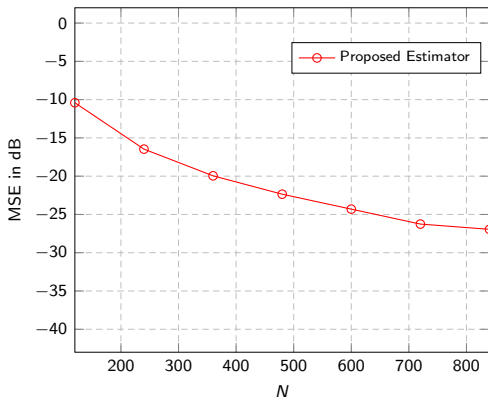
- Consider $\mathbf{s} = -\mathbf{\Gamma}^{-1}\mathbf{b}$, then, the solution $\hat{\beta}_1, \dots, \hat{\beta}_L$ are solutions to the K order polynomial whose coefficients are given by s_0, \dots, s_K .
- The vector $\hat{\mathbf{c}} = \hat{c}_1, \dots, \hat{c}_L$ is given by

$$\hat{\mathbf{c}} = \mathbf{A}^{-1}\mathbf{d}$$

where

$$\mathbf{A} = \begin{bmatrix} 1 & \cdots & 1 \\ \hat{\beta}_1 & \cdots & \hat{\beta}_K \\ \vdots & \vdots & \vdots \\ \hat{\beta}_1^{K-1} & \cdots & \hat{\beta}_K^{K-1} \end{bmatrix} \text{ and } \mathbf{d} = [\hat{\beta}_0, \dots, \hat{\beta}_K]^T$$

Numerical illustrations

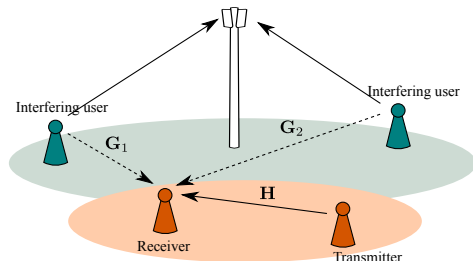


- $\beta_1 = 1, \beta_2 = 2, \beta_3 = 3$
- $\frac{N_1}{N} = 0.5, \frac{N_2}{N} = \frac{N_3}{N} = 0.25,$
- $\frac{N}{n} = \frac{3}{20}$

Consistent estimation of Functionals of the covariance matrix

Work published in [IEEE-IT,2013]

- In some scenarios, the parameter to be estimated is function on the eigenvalues and the eigenvectors of the covariance matrix,
- This scenario cannot be dealt with using the Stieltjes transform method.
- Example of scenario: estimation of the capacity under coloured interference:



Communication model and ergodic capacity

Communication equation: 2 phases

- A learning phase: $\mathbf{Y} = \sum_{k=1}^I \mathbf{G}_k \mathbf{X}_k + \sigma \mathbf{W} \triangleq \mathbf{G} \mathbf{X} + \sigma \mathbf{W}$
- A communication phase: $\mathbf{Y}_2 = \mathbf{H} \mathbf{X}_0 + \sum_{k=1}^I \mathbf{G}_k \mathbf{X}_{k,2} + \sigma \mathbf{W}_2$
- Assumption: The channel \mathbf{H} is perfectly known,

Associated ergodic capacity:

$$C_{\text{erg}} = \frac{1}{N} \log \det (\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H + \mathbf{H} \mathbf{H}^H) - \frac{1}{N} \log \det (\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H)$$

Objective

Estimation of the ergodic capacity:

$$C_{\text{erg}} = \frac{1}{N} \log \det (\sigma^2 \mathbf{I} + \mathbf{G}\mathbf{G}^H + \mathbf{H}\mathbf{H}^H) - \frac{1}{N} \log \det (\sigma^2 \mathbf{I} + \mathbf{G}\mathbf{G}^H)$$

based on the $N \times n$ observations:

$$\mathbf{Y} = \mathbf{G}\mathbf{X} + \sigma\mathbf{W}$$

Regime of interest: n larger but of the same order of magnitude as n . Formally,

$$1 < \liminf \frac{n}{N} \leq \limsup \frac{n}{N} < +\infty.$$

Traditional estimator

Regime when $n \gg N$.

$$\frac{1}{n} \mathbf{Y} \mathbf{Y}^H \xrightarrow[n \rightarrow +\infty, N \text{ fixed}]{} \sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H$$

Hence,

$$\frac{1}{N} \log \det \left(\frac{1}{n} \mathbf{Y} \mathbf{Y}^H + \mathbf{H} \mathbf{H}^H \right) - \frac{1}{N} \log \det \left(\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H + \mathbf{H} \mathbf{H}^H \right) \xrightarrow[n \rightarrow +\infty]{a.s.} 0$$

$$\frac{1}{N} \log \det \left(\frac{1}{n} \mathbf{Y} \mathbf{Y}^H \right) - \frac{1}{N} \log \det \left(\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H \right) \xrightarrow[n \rightarrow +\infty]{a.s.} 0$$

Define the parametrized traditional estimator as:

$$\hat{C}_{\text{trad}}(y) = \frac{1}{N} \log \det \left(\frac{1}{n} \mathbf{Y} \mathbf{Y}^H + y \mathbf{H} \mathbf{H}^H \right) - \frac{1}{N} \log \det \left(\frac{1}{n} \mathbf{Y} \mathbf{Y}^H \right)$$

If N is fixed and $n \rightarrow +\infty$, then:

$$\hat{C}_{\text{trad}}(1) - C_{\text{erg}} \rightarrow 0.$$

Deterministic equivalents

Fundamental equation: Let $y > 0$. The following equation in $\kappa = \kappa(y)$ admits a unique positive solution:

$$\kappa(y) = \frac{1}{n} \operatorname{tr}(\sigma^2 \mathbf{I} + \mathbf{G}\mathbf{G}^H) \left(\frac{\sigma^2 \mathbf{I} + \mathbf{G}\mathbf{G}^H}{1 + \kappa} + y \mathbf{H}\mathbf{H}^H \right)^{-1}$$

admits a unique positive solution,

Auxiliary quantities:

$$\mathbf{T}(y) = \left(y \mathbf{H}\mathbf{H}^H + \frac{\sigma^2 \mathbf{I} + \mathbf{G}\mathbf{G}^H}{1 + \kappa} \right)^{-1}, \quad \mathbf{Q}(y) = \left(y \mathbf{H}\mathbf{H}^H + \frac{1}{n} \mathbf{Y}\mathbf{Y}^H \right)^{-1}.$$

$$\frac{1}{N} \operatorname{tr} \mathbf{Q}(y) - \frac{1}{N} \operatorname{tr} \mathbf{T}(y) \xrightarrow[n, N \rightarrow +\infty]{a.s.} 0.$$

Asymptotic results

For $y > 0$ and also for $y = 0$,

$$\begin{aligned} & \frac{1}{N} \log \det \left(y \mathbf{H} \mathbf{H}^H + \frac{1}{n} \mathbf{Y} \mathbf{Y}^H \right) \\ & - \frac{1}{N} \log \det \left(y \mathbf{H} \mathbf{H}^H + \frac{\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H}{1 + \kappa} \right) - \frac{n}{N} \log(1 + \kappa) + \frac{n}{N} \frac{\kappa}{1 + \kappa} \rightarrow 0. \end{aligned}$$

\Rightarrow , under the regime of interest:

$$\begin{aligned} \hat{\mathcal{C}}_{\text{trad}}(y) & - \left(\frac{1}{N} \log \det \left(y \mathbf{H} \mathbf{H}^H + \frac{\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H}{1 + \kappa} \right) - \frac{1}{N} \log \det (\sigma^2 \mathbf{I} + \mathbf{G} \mathbf{G}^H) \right) \\ & - \frac{n}{N} \log(1 + \kappa) + \frac{n}{N} \frac{\kappa}{1 + \kappa} + \frac{n - N}{N} \log \left(\frac{n - N}{n} \right) - 1 \rightarrow 0. \end{aligned}$$

Setting of y

To make the desired quantity appear in $\hat{C}_{\text{trad}}(y)$, parameter y should be set so that :

$$y = \frac{1}{1 + \kappa(y)}$$

We can prove that there is a unique y^* verifying the above equation. Moreover, a consistent estimate of \hat{y} can be given as the solution to:

$$\hat{y} = 1 - \frac{N}{n} + \frac{\hat{y}}{n} \text{tr} \mathbf{H} \mathbf{H}^H \left(\hat{y} \mathbf{H} \mathbf{H}^H + \frac{1}{n} \mathbf{Y} \mathbf{Y}^H \right)^{-1}$$

Convergence:

$$y^* - \hat{y} \rightarrow 0.$$

Consistent estimator for C_{erg}

A consistent estimator for C_{erg} is given by:

$$\begin{aligned}\widehat{C}_G &= \frac{1}{N} \log \det \left(\frac{1}{n} \mathbf{Y}\mathbf{Y}^H + \widehat{y} \mathbf{H}\mathbf{H}^H \right) - \frac{1}{N} \log \det \left(\frac{1}{n} \mathbf{Y}\mathbf{Y}^H \right) \\ &\quad + \frac{n-N}{N} \left(\log \left(\frac{n\widehat{y}}{n-N} \right) + 1 \right) - \frac{n}{N} \widehat{y}\end{aligned}$$

Fluctuations of the estimator: Let

$$\Theta_N = 2 \log(ny^*) - \log \left((n-N) \left(n - \text{tr} \left(\mathbf{I}_N + \mathbf{H}\mathbf{H}^H (\mathbf{G}\mathbf{G}^H + \sigma^2 \mathbf{I}_N)^{-1} \right)^{-2} \right) \right),$$

Then,

$$0 < \liminf \Theta_N \leq \limsup \Theta_N < +\infty$$

and

$$\frac{N}{\Theta_N} \left(\widehat{C}_G - C_{\text{erg}} \right) \xrightarrow{\mathcal{D}} \mathcal{N}(0, 1)$$

Simulations

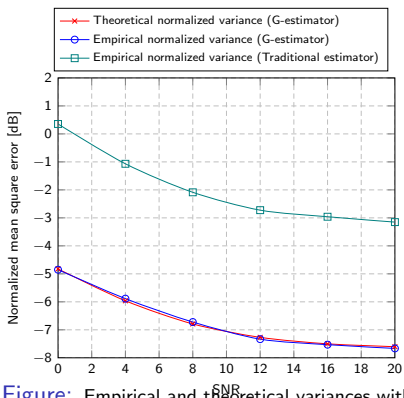


Figure: Empirical and theoretical variances with respect to the SNR

- $N = 4$ %Number of antennas of the receiver
- $n = 15$ %Number of available samples
- $s = 4$ %Number of antennas of the transmitter
- $n = 8$ %Number of interfering mono-antenna users

Simulations

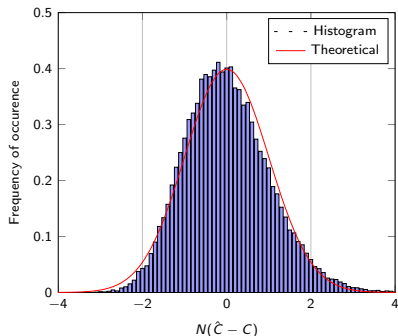


Figure: Histogram of $N(\hat{C} - C)$.

- $N = 4$ %Number of antennas of the interfered receiver
- $n = 15$ %Number of available samples
- $s = 4$ %Number of antennas of the transmitter
- $n = 8$ %Number of the interfering mono-antenna users
- SNR = 10dB

Performance analysis of robust scatter estimates

- In many signal processing applications, the noise is impulsive:
 - ▶ The clutter in radar is in general modeled using elliptical distributions,
 - ▶ Interference in wireless communication can be non Gaussian,
- ⇒ Outliers, even in low numbers, can considerably degrade the performance of usual estimation and detection techniques,
- Commonly used model for impulsive noises, Elliptical distribution:

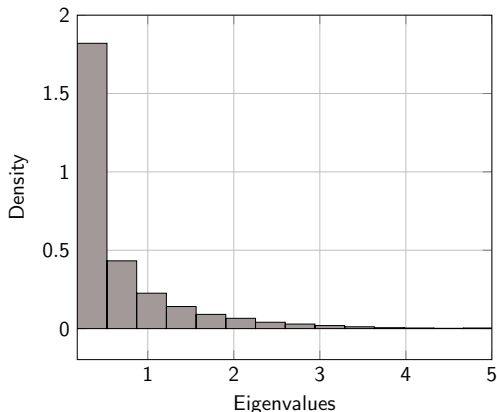
$$\mathbf{x}_i = \sqrt{\tau_i} \mathbf{z}_i \quad \text{où} \quad \begin{cases} \tau_i & \tau_i \text{ heavy tailed.} \\ \mathbf{z}_i & \text{Gaussienne.} \end{cases}$$

Sample covariance matrix under elliptical noises

- Remember that the sample covariance matrix is given by:

$$\hat{\Sigma}_N = \frac{1}{n} \sum_{i=1}^n \tau_i \mathbf{x}_i \mathbf{x}_i^H$$

- The support of the distribution of τ_i being non-compact, The eigenvalue spectrum support will be also unbounded.



Robust scatter matrix estimator

- Let $\mathbf{y}_1, \dots, \mathbf{y}_n$, be n observations of size N and covariance $\mathbf{\Sigma}$. The robust scatter matrix estimator is solution of the following equation:

$$\hat{\mathbf{C}}_N = \frac{1}{n} \sum_{i=1}^n \mathbf{y}_i \mathbf{y}_i^H u\left(\frac{1}{N} \mathbf{y}_i^H \hat{\mathbf{C}}_N^{-1} \mathbf{y}_i\right),$$

where u is a decreasing function such that $\phi = xu(x)$ is increasing.

- Such techniques can be traced back to the works of Huber (1964)
- They have been studied under the regime $N \ll n$.
- Very few results exists in the regime of large random matrices $N \propto n$ (Donoho, ElKaroui, 2013, not yet published).

Performance analysis of robust scatter estimates

- Recent works about the behaviour of the robust scatter methods concern the case where the received observations correspond to a purely noise signal.

[Couillet, JMVA]

$$\mathbf{y}_i = \sqrt{\tau_i} \Sigma^{\frac{1}{2}} \mathbf{z}_i, i = 1, \dots, n$$

- Result:

Theorem

Let $g : \mathbb{R}^+ \rightarrow \mathbb{R}^+, x \mapsto \frac{x}{1 - c_N \phi(x)}$ and $v : x \mapsto (u \circ g^{-1})(x)$, $\psi : x \mapsto xv(x)$. Let γ be the unique positive solution of the equation in γ :

$$1 = \frac{1}{n} \sum_{i=1}^n \frac{\psi(\tau_i \gamma)}{1 + c_N \psi(\tau_i \gamma)}$$

Then:

$$\|\widehat{\mathbf{C}}_N - \widehat{\mathbf{S}}_N\| \xrightarrow{a.s.} 0$$

where

$$\widehat{\mathbf{S}}_N = \frac{1}{n} \sum_{i=1}^n v(\tau_i \gamma_N) \mathbf{y}_i \mathbf{y}_i^H$$

Performance analysis of robust scatter estimates

- Consequence of this result:

$$\hat{\mathbf{S}}_N = \frac{1}{n} \sum_{i=1}^n \tau_i v(\tau_i \gamma_N) \mathbf{z}_i \mathbf{z}_i^H$$

- Function $v(\tau_i \gamma_N) = \frac{\psi(\tau_i \gamma_N)}{\tau_i \gamma_N} \leq \frac{\psi_\infty}{\tau_i \gamma_N}$

$$\hat{\mathbf{S}}_N = \frac{1}{n} \sum_{i=1}^n \frac{\psi(\tau_i \gamma_N)}{\gamma_N} \mathbf{z}_i \mathbf{z}_i^H$$

⇒ The support of the eigenvalue spectrum of $\hat{\mathbf{S}}_N$ is compact.

Elements of the proof

- Define $\hat{\mathbf{C}}_{N,i}$ as:

$$\hat{\mathbf{C}}_{N,i} = \frac{1}{n} \sum_{j=1, j \neq i}^n \mathbf{y}_j \mathbf{y}_j^* u \left(\frac{1}{N} \mathbf{y}_j^H \hat{\mathbf{C}}_{N,i}^{-1} \mathbf{y}_j \right)$$

- Using standard matrix inversion lemma, we can prove that:

$$\frac{1}{N} \mathbf{y}_i^H \hat{\mathbf{C}}_{N,i}^{-1} \mathbf{y}_i = g^{-1} \left(\frac{1}{N} \mathbf{y}_i^H \hat{\mathbf{C}}_{N,i}^{-1} \mathbf{y}_i \right)$$

⇒

$$\hat{\mathbf{C}}_{N,i} = \frac{1}{n} \sum_{j=1, j \neq i}^n \mathbf{y}_j \mathbf{y}_j^* v \left(\frac{1}{N} \mathbf{y}_j^H \hat{\mathbf{C}}_{N,i}^{-1} \mathbf{y}_j \right)$$

- The dependence of $\hat{\mathbf{C}}_{N,i}$ on \mathbf{y}_j for $j \neq i$ is low. Therefore,

$$\frac{1}{N} \mathbf{y}_j^H \hat{\mathbf{C}}_{N,i}^{-1} \mathbf{y}_j \sim \frac{1}{N} \text{tr} \mathbf{\Sigma} \hat{\mathbf{C}}_{N,i}^{-1} \sim \tau_j \gamma_N$$

⇒ In the asymptotic regime, γ_N

$$\gamma_N \sim \frac{1}{N} \text{tr} \left(\frac{1}{n} \sum_{j=1, j \neq i}^n \mathbf{y}_j \mathbf{y}_j^H v(\tau_j \gamma_N) \right)^{-1}$$

Generalization to the signal plus noise

- System model: $\mathbf{y}_i = \mathbf{A}_N \mathbf{s}_i + \mathbf{x}_i$ where $\mathbf{x}_i = \sqrt{\tau_i} \mathbf{z}_i$.
- Result:

Convergence of the robust scatter estimator (work in preparation)

Let $\mathbf{B}_N = \mathbf{A}_N \mathbf{A}_N^*$. Define $\delta_1, \dots, \delta_N$ the unique solutions to the following system of equations:

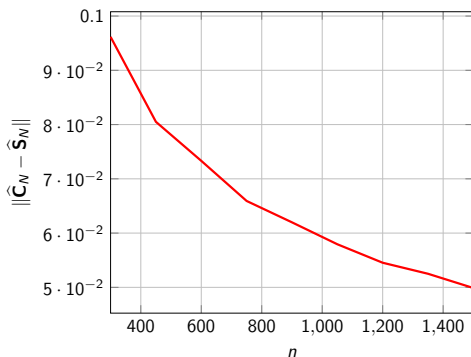
$$\delta_i = \frac{1}{N} \operatorname{tr}(\mathbf{B}_N + \tau_i \mathbf{I}_N) \left(\frac{1}{n} \sum_{j=1}^n \frac{v(\delta_j)(\mathbf{B}_N + \tau_j \mathbf{I}_N)}{1 + c_N \Psi(\delta_j)} \right)^{-1}$$

Then,

$$\|\widehat{\mathbf{C}}_N - \widehat{\mathbf{S}}_N\| \xrightarrow{a.s.} 0$$

where $\widehat{\mathbf{S}}_N = \frac{1}{n} \sum_{i=1}^n v(\delta_i) \mathbf{y}_i \mathbf{y}_i^H$. Moreover, the support of the eigenvalue spectrum of $\widehat{\mathbf{S}}_N$ is compact.

Numerical results



- $t = 1,$
- $u(x) = \frac{1+t}{t+x}$
- $\frac{N}{n} = \frac{1}{3},$

Perspectives

- In case the number of observations is limited, estimates of the covariance matrix are ill-conditioned.
 ⇒ Study of the asymptotic behaviour of the regularized robust scatter matrix:

$$\hat{\mathbf{C}}_N = \frac{1-\rho}{n} \sum_{i=1}^n \mathbf{y}_i \mathbf{y}_i^H u \left(\frac{1}{N} \mathbf{y}_i^H \hat{\mathbf{C}}_N^{-1} \mathbf{y}_i \right) + \rho \mathbf{I}_N$$

- Study of the behaviour of detection methods under elliptical noises, using the results on the information plus noise model
- ⇒ Fluctuations of the eigenvalues of robust scatter estimates of the covariance matrix,
- Study the minimum square error of robust estimation methods,
- ⇒ What is the best function u which achieves the minimum squared error ?